

Cheap Video for Your Heathkit H8

Here's a first look at part of Don Lancaster's latest Sams book, *Son of Cheap Video*. The TV 6-5/8 he talks about is a full graphics update of the original TVT-6L we ran in the June 1977 issue of *Kilobaud*, is a PAIA kit and is detailed in Lancaster's *Cheap Video Cookbook*.

You'll find things more challenging when you add cheap video to an 8080 or Z-80 system, compared to the easy 6500 conversions. There are several new hassles involved that will get in your way and somehow have to be resolved.

In most cases, these hassles will take extra coding, more low-cost ICs and very careful attention to your system timing. The bottom line is this: Cheap video *should* go on most any 8080 or Z-80 system, but it will take more effort, more code and more parts to get comparable results.

Let's see just what these hassles are. First, we'll look at an 8080 in general to see what

the hardware and software differences will be. Then we'll check into a general-use 8080/Z-80 adapter that goes between your computer and the TVT 6-5/8. Finally, we'll show you the software you will need to put cheap video on a Heathkit H8.

We'll assume your system is bus oriented and that your cheap video system is to be a piggyback add-on to an existing RAM plug-in card. We'll further assume the usual 2 MHz 8080 speed. Your RAM should be fast enough that it does not use the READY command to hold up CPU time. We'll also assume your system is big enough that nonvolatile scan firmware

is more important than minimizing the total words of scan coding.

Be forewarned that what we are going to look at has only been tested on the Benton Harbor 50-pin bus. While there is no obvious reason why you can't do the same thing on an S-100 bus system or with a Z-80, we have not tried it just yet, and neither should you... unless you have a good triggered scope on hand and thoroughly understand the 8080 CPU timing.

Our main 8080 hassles are these:

1. The address bus has garbage on it at times.
2. The program counter usually

can change only once every two microseconds. This is only half as fast as we need for a reasonable number of characters or chunks on a line.

3. Clocking and timing signals are different.

4. Literal translation of scan programs will be far too slow.

In general, we'll get around hassle #1 by latching and holding both address and upstream tap data lines using suitably spaced timing. We'll beat #2 by adding a "speed doubling" circuit that creates the *illusion* of a once-per-microsecond program counter advance. This illusion will appear only at the display memory and then only

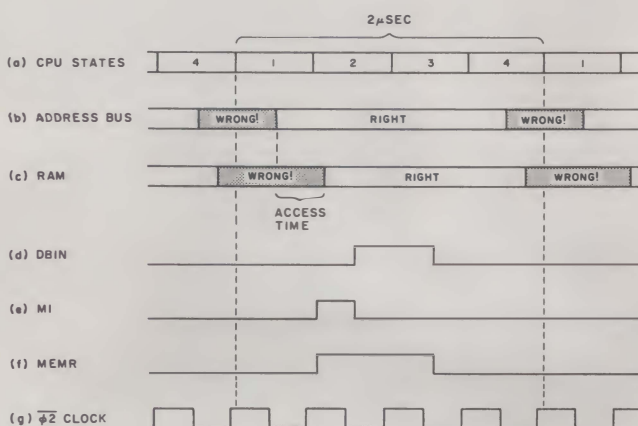


Fig. 1. The H8 is a typical 2 MHz 8080A system. Here are the waveforms involved in reading a NOP command out of RAM.

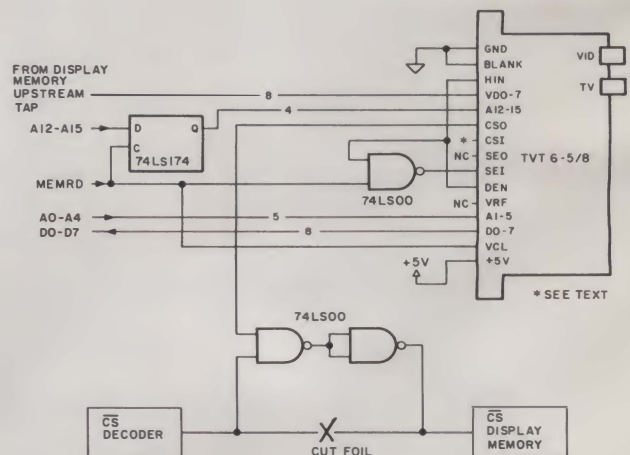


Fig. 2. Minimum 8080A-TV 6-5/8 interface is limited to 2 usec character or chunk times.

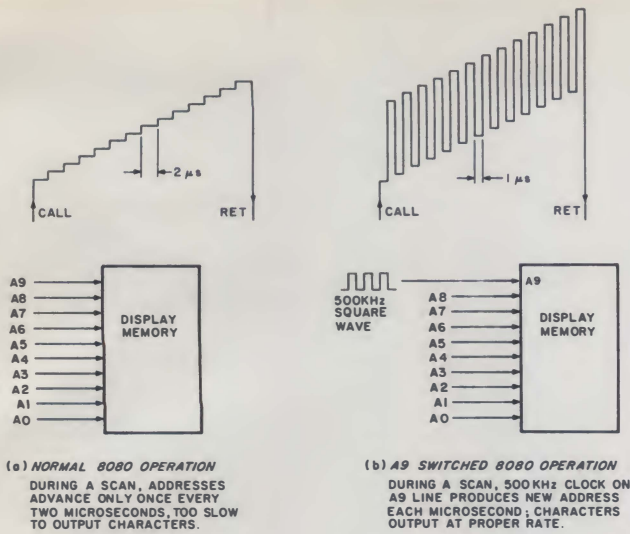


Fig. 3. A stock 8080 system can't change display memory addresses each microsecond. Here's how to use A9 switching for speedup.

during a TVT scan. Everything else stays the usual speed. Hassle #3 goes away when we solve #2. Finally, we can get scan software that is fast enough by using the powerful register-to-register commands of the 8080 or by using brute force (all ROM, non-modifying) coding.

On to the fine print.

Hardware

Suppose we have a normal, functional H8 executing a string of no operations (NOP) from a plug-in RAM card. What will this timing look like? How can we trick the H8 into using the same sort of timing—with additions—to run a TVT 6-5/8? Fig. 1 gives us some clues.

Execution of a NOP takes two microseconds (actually, slightly less than this on the H8). Four CPU states (Fig. 1a), each taking around half a microsecond, are involved. The object of these four states is to put the program counter on the address bus, read an addressed memory location, enter it into the CPU and then act on the command. When the CPU finds out the command is a NOP, it will spend the tail end of the cycle essentially doing nothing.

Our first hassle appears in Fig. 1b. We see that the address bus only has the correct information on it three-quarters of the time. For the remaining

quarter of the time, the address bus has invalid information on it. Now, if we address a memory with the wrong address, we will, of course, get the wrong information out of the memory. Worse still, since the memory has its own access time to contend with, the amount of time that useful information comes out of the memory is even shorter than the time the address bus is valid (Fig. 1c). So, the bad news is that both data and address have all kinds of holes in them and don't seem directly usable.

There are some system-level signals that may help us out of this bind. Signal DBIN in Fig. 1d determines the time when the CPU must have valid data; but this signal is not available on the system bus... for a very good reason. Anyone who tries to use this signal will be cutting into the CPU's own processing time and degrading performance. Instead, two signals are derived for bus use. These signals occur early enough so that enables, decoding, settling times and so on are complete before the CPU needs valid data. These signals are called M1 (Fig. 1e) and MEMR (Fig. 1f).

M1 starts after the address is valid but ends before DBIN. MEMR includes both the M1 and DBIN times. Unfortunately, both M1 and MEMR start before we are sure that the memory is

outputting valid data. The theory here is that output enables and bus access can take place during the same time that the memory is still accessing itself, so long as everything ends up stable by the start of DBIN time. A final waveform we will find useful is the $\overline{O2}$ system clock shown in Fig. 1g.

The least we can get away with and still get cheap video on an 8080 is latching the upper four address lines. If we don't do this, all the commands out of our TVT Instruction decoder PROM, including the row commands and the sync pulses, will have big holes chopped in them.

Fig. 2 shows a minimum 8080-to-TVT 6-5/8 interface. In this circuit, +5, ground, blanking, the upstream tap and the data bus are connected in the usual way. Address lines A12 through A15 are connected to a latch that catches the valid addresses. This is done on the leading edge of the memory read command, MEMR.

Our chip select output CSO is shown going to an AND gate

that gives us an external negative logic OR combination of the old display memory chip select and the one needed for TVT scanning. A foil cut is involved here. The chip select input, CSI, is shown permanently enabled. Depending on your decode PROM, this can go to a TVT enable switch, do nothing or be used as an internal chip select combiner, eliminating the external gate.

The TVT is only allowed to gain data bus control during a scan and then only when the computer wants to read it. To do this, we use the computer's memory read MEMR command and NAND it with the decode PROM, this can go to a TVT enable switch, do nothing or be used as an internal chip select combiner, eliminating the external gate.

MEMR also goes to the clock input of the TVT 6-5/8. But, since our load command in the TVT is derived from the falling edge of VCL, it is the trailing edge of MEMR that loads our video shift register. The time difference of about 750 nanoseconds gives our character generator more than enough time to produce a valid output.

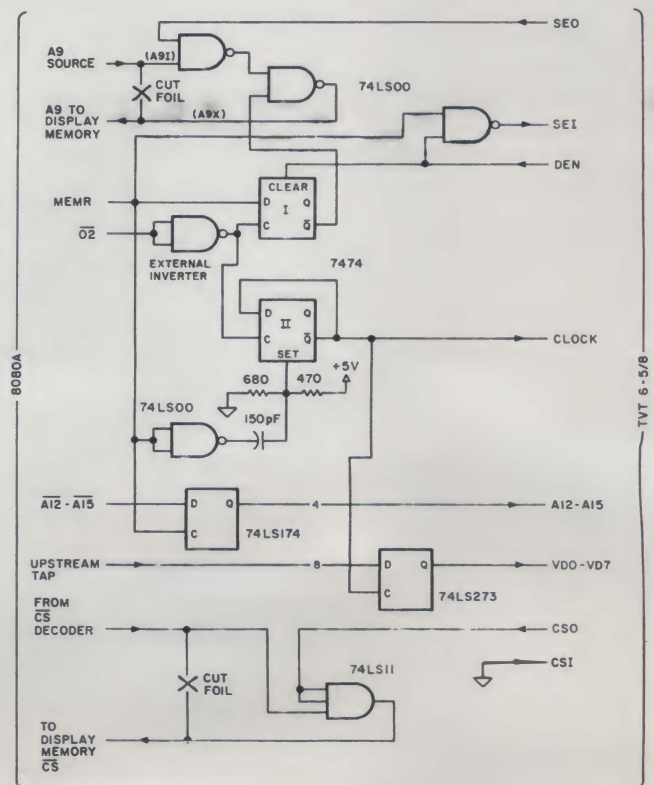


Fig. 4. Speed-doubling 8080A-TV 6-5/8 interface gives 1 usec character or chunk times.

Now, this is a quick and dirty circuit that you may want to try just to get *some* video out of your 8080 in a hurry. But, there are several problems we still have to attack to get something good enough for final system use.

One minor hang-up is that you may only have complements of your data bus or address bus available. We'll soon see how to change the coding in your Scan and Decode PROMs to get around this. The coding, of course, has to be changed anyway since the 8080 gets all bent out of shape when it receives 6502 commands. Inverters or inverting gates can also be used to invert bus, clock, data or control lines as needed.

The big hassle is that the character or chunk times will be *two* microseconds each, rather than just one. This means that, so far, even a 32 character line won't run at normal horizontal scan frequencies. Beating this particular hassle soundly about the head and ears is the key to practical cheap video on the 8080.

But how?

Speed Doubling Via A9 Switching

We want to get our chunk and character times down to a decent rate of one microsecond. We can either speed up the microprocessor or else do something else that creates the *ill-*

lusion of a microprocessor speedup at the display memory and in the adapter circuits.

Speedup may be easy for you if you have a Z-80, provided your display memory is also fast enough to not use the READY command. If you do run faster, you probably would like to latch the upstream tap data to make sure you have enough processing time for your character generator. While a simple speedup will work in some systems, there is a much better way called *A9 switching*.

The object of A9 switching is to create the *illusion* of a once-per-microsecond address advance at the display memory. Fig. 3 gives us details on how this works. We break our most significant display space address line and connect it to a carefully timed 500 kHz square wave during a scan. For a 16 x 64 or a 12 x 80 alphanumeric display, this will be address line A9.

Now, a 500 kHz square wave is low for one microsecond and high for another one. While all the regular addresses *below* A9 are changing at their usual two-microsecond rate, A9 is busy addressing one character or chunk location on the first microsecond and another location on the second. Thus, we get characters or chunks out of our display memory at a one-per-microsecond speed.

But why on earth use A9? Wouldn't it be simpler to use A0 instead? If we do this, we would have to add an address multiplexer to all inputs of the display memory—a 10-pole double throw switch or its Tri-state equivalent. This is obviously something we want to avoid if we are piggybacking video onto an existing memory card. All A9 switching takes a single foil cut and some add-on wires to the memory card.

There is a catch. It is a "yeah-but" rather than a "gotcha." *The characters and chunks are no longer in the display memory in sequential order if you use A9 switching.* So, your cursor or controlling loader software has to have a few words added to complement A9 each successive location.

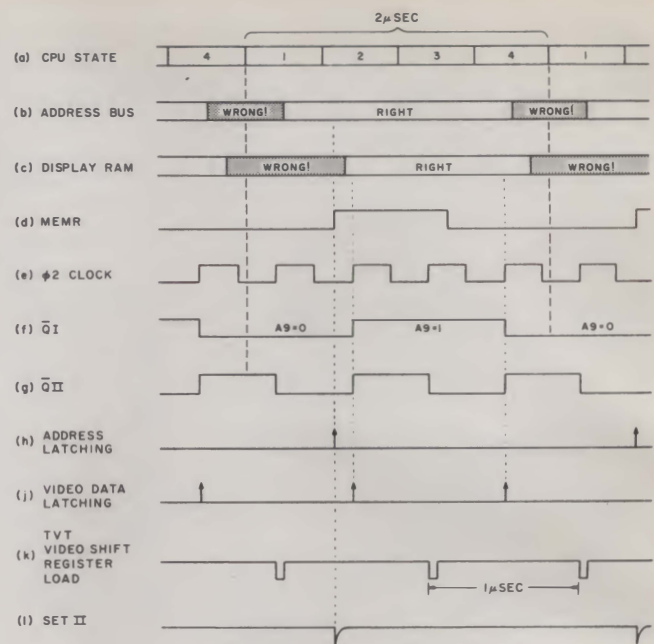


Fig. 5. Speed-doubling waveforms.

For instance, say your display memory starts at 000 000. The next character or chunk will be at 002 000. Your characters will follow in this order:

1st character	000 000
2nd character	002 000
3rd character	000 001
4th character	002 001
5th character	000 002
6th character	002 002
.	.
1022nd character	003 376
1023rd character	001 377
1024th character	003 377

This seems awful, but it works. And it is a simple way to double the apparent memory access speed of an 8080 so we can get information out of RAM once per microsecond under block access. And all it takes to do the job is some simple hardware between computer and TVT, a few software words and one extra foil cut on the memory. The hardware involved is shown in Fig. 4, along with the timing details of Fig. 5.

Two new D-flip-flops are added to our interface. The first delays and expands the MEMR signal to give us a controlled phase 500 kHz square wave we can use for the speed doubling A9 address switching. The second divides the system clock by two and is used to latch the video data and to provide a TVT clock.

Waveforms (a), (b), (c) and (d)

in Fig. 5 are as before. Waveform (e) is a $\phi 2$ clock, which has to be an inverted replica of the Heath bus $\phi 2$ clock signal. Waveform (f) shows us the 500 kHz square wave that results when we clock MEMR. Since the clocking is delayed from the MEMR leading edge, the flip-flop's output is wider than MEMR and is almost a microsecond long. This results in a square wave that is low for one microsecond and high for the next, locked to (but following) MEMR.

This particular flip-flop is only allowed to run *during* a scan. Otherwise, it is held high by DEN. The uppermost two gates combine the old A9 information with the speed-doubling new A9 signal, acting as a single pole, double throw selector switch. During computer times, the display memory A9 line is connected to the computer. During scan microinstruction times, the display memory A9 line is connected so it is low for one microsecond and high for the next.

Waveform (g) shows us the one megahertz clock we get by dividing down $\phi 2$. This clock is used to sample and latch the display memory output immediately after the data is valid and then latch again one microsecond later, well after the A9

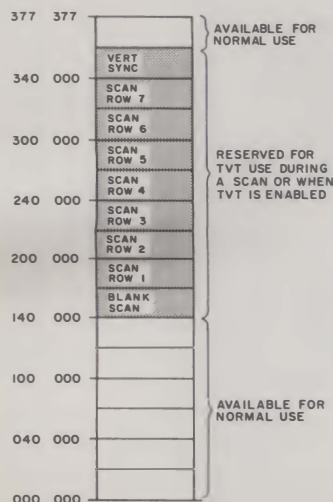


Fig. 6. H8 address map.

658-HD8

PROM NUMBER

□ = "0" ■ = "1" (POSITIVE LOGIC)

USE FOR TVT 6-5/8 ON AN 8080 SYSTEM WITH INVERTED A12, A13, A14, A15 LINES.

CG LINE "2" IS USED AS GRAPHICS "BLANKING" OUTPUT.

CG LINE "4" IS USED AS GRAPHICS "UPPER-LOWER" CHUNK SELECT OUTPUT.

WORD #	WHAT DOES THIS WORD DO?	HEX OP-CODE	INPUTS								OUTPUTS							
			CS OUT	Q8	Q7	Q6	Q5	Q4	Q3	Q2	Q1	SCAN ENABLE	DECODE ENABLE	VERTICAL SYNC	(SPARE)	CG LINE "4"	CG LINE "2"	CG LINE "1"
0	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
1	VERTICAL SYNC	d0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
2	LINE 7 SCAN	27	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
3	LINE 6 SCAN	26	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
4	LINE 5 SCAN	25	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
5	LINE 4 SCAN	24	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
6	LINE 3 SCAN	23	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
7	LINE 2 SCAN	22	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
8	LINE 1 SCAN	21	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
9	BLANK SCAN	20	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
10	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
11	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
12	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
13	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
14	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
15	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
16	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
17	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
18	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
19	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
20	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
21	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
22	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
23	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
24	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
25	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
26	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
27	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
28	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
29	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
30	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
31	NORMAL	C0	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■

Fig. 7. Truth table for 8080 Decode PROM having inverted address inputs (used on Heathkit H8).

change has been accepted. The first sample gives us an A9=0 data value, while the second handles the A9=1 case. The TVT's video shift register is clocked on the falling edge of this one megahertz clock. Since there is a one-half microsecond delay between the leading and trailing clock edges, enough time is available for the character generator or the data-to-video converter to accept the latched video data and process it.

Our A9-generating flip-flop automatically initializes itself on MEMR since it is simply delaying this signal. But the clock-dividing flip-flop can be in either state at the beginning of a scan microinstruction. Unless we somehow initialize this flip-flop to the right state, we'll get garbage out of the display memory caused by sampling at the

wrong times.

We initialize this clock-dividing flip-flop by inverting MEMR and using the leading edge to SET the divide flip-flop to the desired state. This initialization is very important since the usual CALL instruction preceding the scan microinstruction has an odd number of clock cycles in it.

TVT scan enabling and the display memory chip selecting are done the same way as the slower interface of Fig. 2. We enable the TVT Scan Enable Input (SEI) only during MEMR time to give us data for a scan microinstruction only when it is called for and only when the computer will allow data bus access. The display memory chip select is a negative logic OR of the computer's chip select and the CS0 that the TVT provides.

Our speed doubling interface takes two foil cuts on the memory board—one on the A9 address line and one on the chip select line. All other connections are add-ons derived from signals available on a typical plug-in memory card. Five low-cost integrated circuits are involved in this particular adapter.

Software

Let's take a look at the PROM firmware and some of the software involved in getting cheap video on your 8080A system. For right now, we'll stick to the older address-mapped and sub-routine-scanned methods we used in the Cheap Video Cookbook. Most likely you can simplify things a great deal by going to the Scungy Video route of break-mapping and interrupt-scanning. The strong Input/Output commands in the 8080A make this a very attractive idea.

If you use address mapping,

refer to the computer memory map shown in Fig. 6. A block of addresses from 6K to 60K is reserved for TVT use when the TVT is enabled. On the H8, this leaves the bottom 8K for the PAM monitor and operating system and 16K for enough RAM to hold a display memory and run Extended BASIC at the same time. The uppermost 4K of addresses are also available as needed.

Should you want more address space for other uses, you can use the TVT enable to free addresses during non-display times. You can also go the Scungy Video route and use I/O instructions and a parallel port instead of address mapping the row commands. Yet another alternative is to use further decoding to activate the TVT only during valid display memory addresses. For instance, if you are only using 1K of display memory, 3K of all the scan blocks can be used for other

*Scungy Video is an alternate method and is detailed in Chapter 1 of Son of Cheap Video.

658-HS64

PROM NUMBER

□ = "0" ■ = "1" (POSITIVE LOGIC)

USE FOR TVT 6-5/8 ON AN 8080 SYSTEM WITH TRUE A0-A7 LINES AND INVERTED DATA BUS NO REPACKING

WORD #	WHAT DOES THIS WORD DO?	HEX OP-CODE	INPUTS								OUTPUTS							
			DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0	Q8	Q7	Q6	Q5	Q4	Q3	Q2	Q1
0	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
1	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
2	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
3	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
4	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
5	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
6	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
7	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
8	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
9	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
10	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
11	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
12	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
13	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
14	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
15	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
16	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
17	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
18	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
19	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
20	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
21	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
22	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
23	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
24	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
25	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
26	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
27	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
28	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
29	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
30	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
31	RET	36	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■

Fig. 8. Truth table for 8080 Scan PROM having no repacking, true address inputs and inverted data outputs.

USE ONLY FOR 80 CHARACTER REPACKED LINES ON AN 8080 SYSTEM WITH TRUE A0-A7 LINES AND INVERTED DATA BUS.

WORD #	WHAT DOES THIS WORD DO?	HEX OP-CODE	OUTPUTS																	
			Q8	Q7	Q6	Q5	Q4	Q3	Q2	Q1										
			DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0										
0	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
1	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
2	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
3	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
4	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
5	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
6	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
7	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
8	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
9	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
10	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
11	RET	36	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□
12	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
13	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
14	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
15	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
16	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
17	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
18	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
19	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
20	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
21	RET	36	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□
22	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
23	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
24	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
25	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
26	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
27	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
28	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
29	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
30	NOP	FF	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■	■
31	RET	36	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□	□

Fig. 9. Truth table for 80 character 8080 Scan PROM (true address inputs, inverted data outputs).

purposes if you add suitable decoding.

A quick look at the H8-3 memory board shows that only some of the address and data lines are available in their true form; most of them are inverted. The data-out buffer on this memory card must be disabled for the upstream tap needed by cheap video. This means that the output of our Scan Microinstruction PROM has to directly drive the system data bus and thus must output inverted (negative logic) data. We also see that address lines A13, A14 and A15 aren't available except as complements. The simplest way out of this situation is to code our Decode PROM to respond directly to complemented addresses.

Fig. 7 shows us the H8 Decode PROM truth table, 658-HD8. We input lines $\bar{A}12$, $\bar{A}13$, $\bar{A}14$ and $\bar{A}15$, along with a TVT enable using the old CSI line. This PROM outputs code to the

row commands of the character generator or else routes blanking and selection commands to a graphics data-to-video converter. The Decode PROM also outputs system controlling signals DEN, SEO, CSO and the vertical sync VRF pulses.

Since we are using complemented address inputs, this PROM runs "backwards" from the earlier PROMs we looked at. The net result of a "frontwards" PROM with true address inputs or a "backwards" PROM with inverted address inputs is the same.

Holding the CSI line positive disables the TVT and frees most all addresses for other uses. Grounding CSI enables the TVT scanning and reserves the needed address blocks for TVT use. This particular PROM coding needs an external AND gate for chip selection and combination.

There are two types of Scan

PROM coding we might like to use, depending on whether we are using "binary" line lengths or are repacking "non-binary" line lengths for maximum memory efficiency. Fig. 8 shows a Scan PROM coding intended for 64 character lines, but usable for 32 character lines, most graphics and other lengths *without* memory repacking. This is numbered 658-HS64.

We use a NOP to advance the program counter in the computer and an RET coding to return from the called scan microinstruction. Since we are outputting complemented data, these outputs are inverted. On the H8, address lines A0 through A6 are available in true form, so we do not have to complement the address inputs. Thus, our Scan PROMs run "frontwards" but output complemented code.

We can use the 658-HS80 Scan PROM truth table in Fig. 9 for memory repacked scans of 80 characters per line, three lines per page. Once again, this PROM coding is driven by true addresses and outputs complementary data directly to the H8 data bus.

Our address lines are connected differently on an 8080 system than on a 6502. Remember that we used every *second* address change on the 6502 to advance our Scan PROM *one* count. On an 8080 we use *every* address change to advance the Scan PROM *one* count, but use A9 switching to get two characters out of memory per one Scan PROM count advance. Either way, the Scan PROM responds to an input address

change once every two micro-seconds, and everything comes out even.

This means that, in general on an 8080 system, the Scan PROM's inputs are usually connected to one address line *less* than usual for a 6502 system. Fig. 10 shows our address line management for an 8080 adapter. It also shows how two new switches can be added along with a gate to let you use either a 658-HS64 or a 658-HS80 Scan PROM on an 8080 system without needing any rewiring.

Several examples will show how this address management works.

1. For 32 character lines using speed doubling, use PROM 658-HS64 and set your switches to A4 = "+", A5 = "+" and "32."
2. For 64 character lines using speed doubling, use PROM 658-HS64 and set your switches to A4 = "A4," A5 = "+" and "32."
3. For 80 character lines using speed doubling and memory repacking, use PROM 658-HS80 and set your switches to A4 = "A4," A5 = "A5" and "64."

In our first example, the upper half of a Scan PROM is cycled through in 16 counts lasting 32 microseconds. In the second example, the entire Scan PROM is cycled through in 32 counts lasting 64 microseconds. In the final example, if we wanted to, the entire Scan PROM could be scanned in 32 counts lasting 256 microseconds. But with memory repacking and A9 switching, we only use slightly under a third of the 80 line Scan PROM *per scan*, ending up with ten counts per scan lasting 80 microseconds.

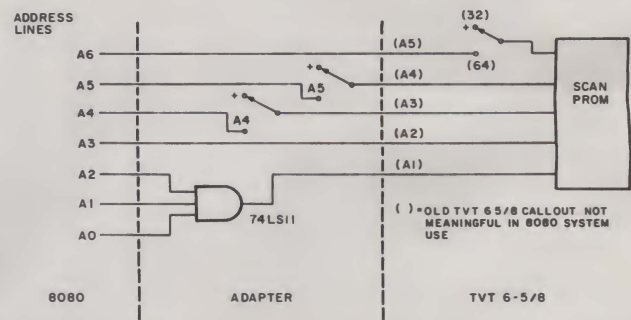


Fig. 10. The Scan PROM address inputs on the TVT 6-5/8 have to be redefined for 8080 use. The gate and switches let you run ordinary or repacked memory PROMs without wire changes.

Your turn: Show the Scan PROM truth table and switch settings for an H8 Scan of 40 repacked characters per line.

Front Panel Interaction

The H8 front panel works by interrupting a running program

once every two milliseconds. If we try to run scan software and the front panel at the same time, the display will be badly torn up. So, we can either turn the front panel off during display times or else combine the front panel and the video scan

into a single program. Just turning the front panel off is far simpler and usually all you will need to do.

The H8 front panel monitor does have a "turn the display off" software word. But this won't help us. While this command shortens the interrupt and keeps it from lighting the display, the interrupt still exists.

One hardware solution is shown in Fig. 11. A new switch is added to the front panel that prevents timer-generated level 10 interrupts from happening. This, in turn, keeps the panel display off and the video display in one piece. This switch will be very handy during your initial test and debugging of video displays. You should only turn off the front panel after you have a video display, and turn it back on before returning to other uses. The RST/0 command does bypass this switch so that you can reset under any conditions.

This switch will most likely not be needed when your properly designed and debugged scan software is operational. You probably can eliminate it from the final use circuitry.

The obvious question is how to use software instead. We have a good old DI, or "disable interrupts," command in the 8080 instruction set. Can't we simply use this?

Unfortunately, there is one very noisy gotcha that may keep you from doing this—unless you are careful.

If you try an immediate DI command in an H8 program, the speaker will latch on and stay on. That little beep you get when you hit the GO key—or any other key—needs two more interrupts after your program starts. No interrupts, no stopping. The two interrupts time out a four millisecond tic for the horn circuit.

The H8 front panel monitor needs a few milliseconds after it is exited before you can disable any interrupts. If you disable an interrupt too soon you will lock the speaker on.

You can use the DI command to turn off the front panel, but you must delay at least five milliseconds after your program

starts or the speaker won't quit. Thus, one properly placed software word is all you need to get full front panel and video display compatibility.

Test Software

Two useful test routines are shown in Fig. 12. Fig. 12a checks Scan PROM access and operation. If this test fails, you are either incorrectly picking up scan microinstructions or are missing them entirely. Erratic switching between 311 (return) and 000 (no operation) means you have speed-doubling problems. All 000s means you are never activating the Scan PROM, while all 311s means you are permanently trying to return from a Scan Microinstruction call. This particular test works with either HS64 or HS80 Scan PROMs and can have the address switches in any position.

Your turn: Why?

Don't ever try going beyond this test if the test fails. If you cannot read the proper return from a scan microinstruction, it will not execute, and anything else you add in the way of software or time or effort will only compound the felony.

Test sequence Fig. 12b lets you transfer control of the H8 from computer to TVT scanning and back again. Note that the test coding differs for each Scan PROM and that each Scan PROM has to have the address switches set as shown.

The scanning process is adjusted to output a TV horizontal scan at normal scan frequencies. In a completely working system with a disabled front panel, you'll get a continuous series of vertical stripes. This corresponds to the seventh dot row of a random character load. A wildly wrong horizontal scan frequency usually means the wrong switch settings or the wrong Scan PROM. Vertical stripes that have teeth in them may be caused by erratic data latching or improper speed-doubling operation.

While these two tests appear trivially simple, don't overlook them as major debugging aids. If these two won't go, no other software will run either.

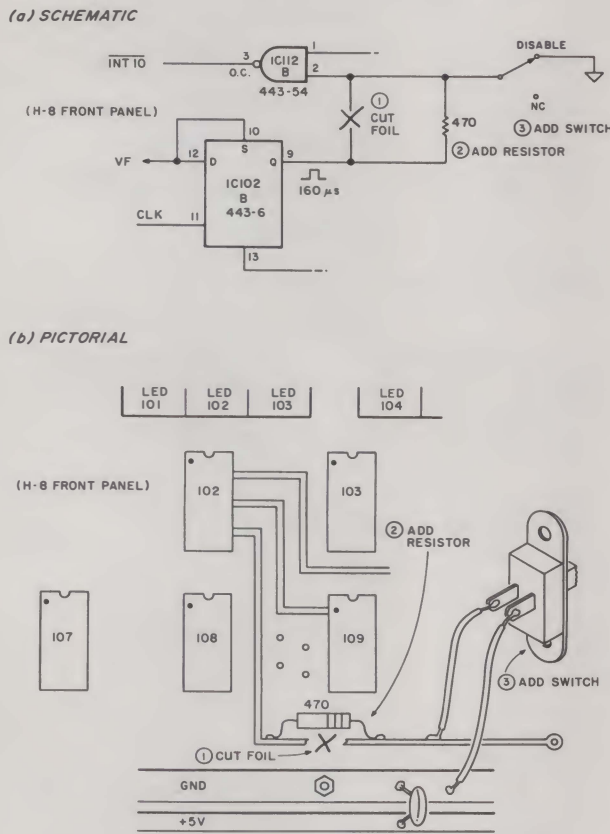


Fig. 11. A switch to temporarily defeat the H8 front panel display will be useful for TVT debugging and checkout.

A. To verify that the Scan Microinstruction is alive and well:

read

300 376	for	000	(NOP)
300 377	for	311	(RET)
301 000	for	000	(NOP)

Either the HS64 or the HS80 Scan PROM may be used.
The address switches may be in any position.

B. To pass control to and from the Scan Microinstruction at a TV Horizontal rate:

For Scan PROM HS64

Set switches to "32"; A5 = "+" and A4 = "A4"

START	→ 040 100	CALL	315 010 320	Scan seventh dot row
	→ 040 103	JMP	303 100 040	Repeat

For Scan PROM HS80

Set switches to "64"; A5 = "A5" and A4 = "A4"

START	→ 040 100	CALL	315 030 320	Scan seventh dot row
	→ 040 103	JMP	303 100 040	Repeat

This will display continuous vertical stripes that correspond to the seventh dot row of a random character load. The front panel should be switch disabled during viewing times.

H8 Scan time is 63 microseconds for a horizontal scan frequency of 15.898 kHz. There is no vertical sync.

Fig. 12. Two test routines useful in 8080/TVT debugging.

Self-Modifying vs Brute Force Scans

The obvious next thing to do is take the old 6502 scan software programs and literally translate them, replacing a CALL for a JSR and so on. But we really get into trouble in a hurry if we try this. First, some commands will be longer or shorter than their 6502 counterparts, messing up the critical horizontal-edge-to-horizontal-edge timing. Worse yet, the execution time of an 8080 working with literally translated 6502 commands is pitifully slow—so slow that the critical timing loop may take over 30 microseconds, compared to the 21 used in the 6502. This makes the long horizontal lines so long we don't want to even think about using them.

One solution is to make the 8080 into an 8080 system rather than an imitation 6502. You can do this using the fast register-to-register transfer commands and get your loop times down only slightly longer than those in the 6502 programs.

But is this really what we want in an 8080 system? Remember that on a bare-bones KIM-1 our back was to the wall in finding room for a scan program. We *had* to get by with the absolute minimum-length scan programs in order to get any video at all.

One result of this restriction was that our scan code was *self-modifying*. This meant that the scan program computed its next set of memory locations rather than looking them up. This, in turn, meant that the scan program *had* to be in RAM during final operation, at least on a KIM.

Usually our 8080 systems have enough RAM and PROM available that we needn't worry too much about minimizing code. So, why not use *brute force* coding that calls each scan address as it is needed? We can store the whole scan program in ROM or PROM this way and never have to load it again... or worry about it bombing when something bad happens in RAM.

Brute force coding will also be much faster. It will be much

easier to write, modify and debug. But, as usual, there is a price. Brute force coding can be much longer than self-modifying coding. On a one-line display, this turns out to be a no-hassle 43 words versus the 30 words we needed on a KIM with self-modifying code. But on a long and involved program such as a 24 x 80 double-stuffed scan, it could take 600 or more words of code to get us by. Still, that's only little over half a 2708 or slightly over a quarter of a 2716 EPROM and no real big deal these days.

Let's use this brute force approach to generate a simple one line display and then apply it to a 12 x 80 scan program.

1 x 56 Scan Program

Fig. 13 shows a brute force scan program for a one line, 56 character no-interlace 8080/TVT 6-5/8 display. Each successive dot row is called by a scan subroutine as it is needed. We start in 040 100 with a short blank scan to get us off on the right

foot. Then we sequentially call dot rows 1 through 7 of the characters to be displayed. This live scanning is followed by a vertical sync pulse.

After this, a word that sets the number of blank scans is loaded in the accumulator (365). As many blank scans as needed are generated in turn. Each time a blank scan is completed, the accumulator word is decremented till the word hits zero. At that time, the program jumps to the top line blank scan and

repeats for the next field.

Unlike a 6502, an 8080 can take an even or an odd number of *half* microseconds to complete an instruction. In most scan programs, some equalization will be needed to make up for this half-microsecond jitter. The command MOVAA, or "move the accumulator to itself," takes 2.5 microseconds and is a benign instruction. This lets us shift timing by half a microsecond if used once and by one microsecond if used

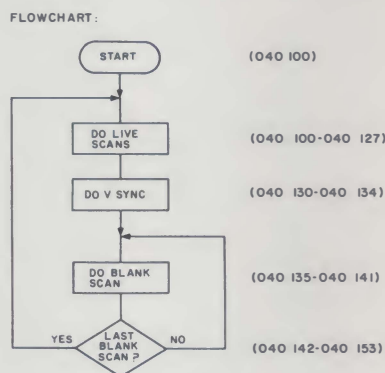


Fig. 13a. Program flowchart.

uP-8080A System-H8		Start-JMP 040 100 End-RST/0	Displayed	340 004 to 340 037 342 004 to 342 037
			Program Space	040 100 to 040 152 (43 words)
START	040 100	CALL 315 017 140		Do short blank scan
	040 103	CALL 315 004 160		Scan Dot row #1
	040 106	CALL 315 004 200		Scan Dot row #2
	040 111	CALL 315 004 220		Scan Dot row #3
	040 114	CALL 315 004 240		Scan Dot row #4
	040 117	CALL 315 004 260		Scan Dot row #5
	040 122	CALL 315 004 300		Scan Dot row #6
	040 125	CALL 315 004 320		Scan Dot row #7
	040 130	LDA 072 000 340		Output Vertical sync pulse
	040 133	MVIA 076 365		Load # of blank scans
	040 135	CALL 315 011 140		Do blank scan
	040 140	DCRA 075		One less scan
	040 141	MOVAA 177		Equalize 2.5 microseconds
	040 142	JNZ 302 (135) (040)		One more blank scan?
	040 145	MOVAA 177		Equalize 5.0 microseconds
	040 146	MOVAA 177		continued
	040 147	DI 363		Shut off horn
	040 150	JMP 303 (100) (040)		Go to live scans

Mods:
 To relocate display space, use program jumpers on memory card or else change starting address of dot scans.
 To put both halves of display space closer together, use A4 switching rather than A9 switching.
 For double height characters, repeat scan of each dot row twice.

Notes:
 ● TVT 6-5/8 must be connected via an 8080 adapter, and both the 658-HD8 and 658-HS64 PROMs must be in circuit for the program to run.
 ● Horizontal frequency 15.174 kHz; Vertical frequency 59.976 Hz. 2500 second hum bar.
 ● Address switches must be in "32", A5 = "+", and A4 = "A4" positions.
 ● Character sequence goes 340 004; 342 004; 340 005; 342 005; 340 006; 342 006; 340 007.
 ● () denotes an absolute address that is program location sensitive.
 ● This program is not self-modifying and may be placed in PROM or ROM.

Fig. 13 Program for a one line, 56-character, no-interlace TVT 6-5/8 8080 raster scan.

Fig. 14. Program for a 12 line, 80-character-per-line, full-interlace, double-stuffed TVT 6-5/8 raster scan.

	uP-8080A	Start-RUN 040 100	
	System-H8	End-RST/0	
	Displayed	340 010 to 343 377	
	Program Space	040 100 to 042 007 (455 words)	

		(even field)	
START →	040 100	CALL 315 023 140	Do short blank scan
	040 103	CALL 315 010 140	Scan dot row 0, character line 1
	040 106	CALL 315 010 200	" 2 " 1
	040 111	CALL 315 010 240	" 4 " 1
	040 114	CALL 315 010 300	" 6 " 1
	040 117	CALL 315 010 140	Do blank scan
	040 122	CALL 315 060 140	Scan dot row 0, character line 2
	040 125	CALL 315 060 200	" 2 " 2
	040 130	CALL 315 060 240	" 4 " 2
	040 133	CALL 315 060 300	" 6 " 2
	040 136	CALL 315 060 140	Do blank scan
	040 141	CALL 315 130 140	Scan dot row 0, character line 3
	040 144	CALL 315 130 200	" 2 " 3
	040 147	CALL 315 130 240	" 4 " 3
	040 152	CALL 315 130 300	" 6 " 3
	040 155	CALL 315 130 140	Do blank scan
	040 160	CALL 315 210 140	Scan dot row 0, character line 4
	040 163	CALL 315 210 200	" 2 " 4
	040 166	CALL 315 210 240	" 4 " 4
	040 171	CALL 315 210 300	" 6 " 4
	040 174	CALL 315 210 140	Do blank scan
	040 177	CALL 315 260 140	Scan dot row 0, character line 5
	040 202	CALL 315 260 200	" 2 " 5
	040 205	CALL 315 260 240	" 4 " 5
	040 210	CALL 315 260 300	" 6 " 5
	040 213	CALL 315 260 140	Do blank scan
	040 216	CALL 315 330 140	Scan dot row 0, character line 6
	040 221	CALL 315 330 200	" 2 " 6
	040 224	CALL 315 330 240	" 4 " 6
	040 227	CALL 315 330 300	" 6 " 6
	040 232	CALL 315 330 140	Do blank scan
	040 235	CALL 315 010 141	Scan dot row 0, character line 7
	040 240	CALL 315 010 201	" 2 " 7
	040 243	CALL 315 010 241	" 4 " 7
	040 246	CALL 315 010 301	" 6 " 7
	040 251	CALL 315 010 141	Do blank scan
	040 254	CALL 315 060 141	Scan dot row 0, character line 8
	040 257	CALL 315 060 201	" 2 " 8
	040 262	CALL 315 060 241	" 4 " 8
	040 265	CALL 315 060 301	" 6 " 8
	040 270	CALL 315 060 141	Do blank scan
	040 273	CALL 315 130 141	Scan dot row 0, character line 9
	040 276	CALL 315 130 201	" 2 " 9
	040 301	CALL 315 130 241	" 4 " 9
	040 304	CALL 315 130 301	" 6 " 9
	040 307	CALL 315 130 141	Do blank scan
	040 312	CALL 315 210 141	Scan dot row 0, character line 10
	040 315	CALL 315 210 201	" 2 " 10
	040 320	CALL 315 210 241	" 4 " 10
	040 323	CALL 315 210 301	" 6 " 10
	040 326	CALL 315 210 141	Do blank scan
	040 331	CALL 315 260 141	Scan dot row 0, character line 11
	040 334	CALL 315 260 201	" 2 " 11
	040 337	CALL 315 260 241	" 4 " 11
	040 342	CALL 315 260 301	" 6 " 11
	040 345	CALL 315 260 141	Do blank scan
	040 350	CALL 315 330 141	Scan dot row 0, character line 12
	040 353	CALL 315 330 201	" 2 " 12
	040 356	CALL 315 330 241	" 4 " 12
	040 361	CALL 315 330 301	" 6 " 12
	040 364	CALL 315 330 141	Do blank scan
	040 367	MVIA 076 006	Delay 48.5 microseconds
	040 371	DCRA 075	continued
	040 372	JNZ 302 (371)(040)	continued
	040 375	LDA 072 000 340	Output //VERTICAL SYNC// pulse
	041 000	CALL 315 363 140	Do short blank scan
	041 003	LDA 072 000 000	Delay 6.5 microseconds
	041 006	MVIA 076 175	Load # of vertical blank scans
	041 010	CALL 315 015 140	Do //BLANK VERTICAL SCANS//
	041 013	DCRA 075	One less blank scan
	041 014	MOVAA 177	Equalize 2.5 microseconds
	041 015	JNZ 302 (010)(041)	Repeat blank scans if not done
	041 020	MOVAA 177	Equalize 5 microseconds
	041 021	MOVAA 177	continued
	041 022	DI 363	Shut off horn

twice. This is the purpose of those strange "177" instructions in the program.

In step 040 147, we disable the interrupts. This turns off our front panel but does so late enough that we will not lock the speaker on. Since the code is not self-modifying, you can put it in your choice of RAM, ROM, PROM, EPROM or E²PROM. Naturally, you'll want to check things out in RAM first before committing yourself to permanent code.

Your turn: Show the coding needed for 1×32, 1×64 and 1×80 scans.

As a hint that will save you lots of trial and error or calculations, keep your blank initial scan *nine* counts short of the live scans and keep the retrace blank scans *five* counts short of your live scans. A stationary or near-stationary hum bar is picked up by adjusting 040 134 as needed. A more obvious route to shorter scans is to simply use the 1×56 and load blanks as needed in unused character locations.

TV Retrace Hassles

Calling and returning from a subroutine takes around 13.5 microseconds on a typical 8080. Two of these microseconds are spent on the live scan, leaving us with a retrace time of 11.5 microseconds. Since the H8 is slightly faster than this, our available retrace time is around 11.2 microseconds.

Naturally, we would like to keep our retrace times as short as possible. This lets you put more characters on the line for standard horizontal rates or lets you run long character lines with more nearly normal horizontal frequencies.

But 11 microseconds may not be enough time for your monitor or TV set to cleanly get from the end of one line to the beginning of the next. For most monitors and some TV sets, 11 microseconds will be just barely enough.

If you are having trouble displaying all the characters, here are some hints that may help you:

● Your simplest out is to adjust

the display centering so that the first character is always legible. Always stop short of the maximum display length on your statements.

● Use the *maximum* possible width. Raising the width coil inductance (see *Cheap Video Cookbook*, Fig. 3-33) can lengthen the needed retrace time.

● Use a longer-than-needed character line and put permanent blanks where they are called for.

● Add equalization to lengthen each CALL sequence. While this is the obvious and cleanest route, it can add many words to a brute force scan program.

● If you *thoroughly* understand TV horizontal scanning and have a decent scope and full TV documentation, modify the fly-back capacitor value as needed to get a faster retrace. But, be careful to not exceed the peak allowable horizontal output transistor voltage when you do this.

More Characters

Our 1×56 scan has several obvious limitations. From this starting point, we'll want to add interlace, double stuffing and lots more characters.

The optimum number of characters or chunks per line seems to be 56 for an H8 system using A9 switching for speed doubling. This 56-character length lets you use a standard horizontal frequency. You can display on either a color or a black and white set.

But there seems to be something magical about 80 character lines that appeals to people, even though this many characters are hard to read and are rarely, if ever, needed. So, to prove it can be done, we're going to show you how to display 80 character lines on your H8 and then put those lines on a TV with unmodified video bandwidth or over an rf modulator. Remember, though, that we'll have to run at a reduced horizontal rate, which will take width and hold modifications to your small-screen, transformer-operated, *Photofact*-available, black and white set. Furthermore, your wrong choice

041 023	JMP	303	(100)(041)	Jump to odd field
(041 026 to 041 077 are spares)				
(odd field)				
041 100	CALL	315	023 140	Do short blank scan
041 103	CALL	315	010 160	Scan dot row 1, character line 1
041 106	CALL	315	010 220	" 3 " 1
041 111	CALL	315	010 260	" 5 " 1
041 114	CALL	315	010 320	" 7 " 1
041 117	CALL	315	010 140	Do blank scan
041 122	CALL	315	060 160	Scan dot row 1, character line 2
041 125	CALL	315	060 220	" 3 " 2
041 130	CALL	315	060 260	" 5 " 2
041 133	CALL	315	060 320	" 7 " 2
041 136	CALL	315	060 140	Do blank scan
041 141	CALL	315	130 160	Scan dot row 1, character line 3
041 144	CALL	315	130 220	" 3 " 3
041 147	CALL	315	130 260	" 5 " 3
041 152	CALL	315	130 320	" 7 " 3
041 155	CALL	315	130 140	Do blank scan
041 160	CALL	315	210 160	Scan dot row 1, character line 4
041 163	CALL	315	210 220	" 3 " 4
041 166	CALL	315	210 260	" 5 " 4
041 171	CALL	315	210 320	" 7 " 4
041 174	CALL	315	210 140	Do blank scan
041 177	CALL	315	260 160	Scan dot row 1, character line 5
041 202	CALL	315	260 220	" 3 " 5
041 205	CALL	315	260 260	" 5 " 5
041 210	CALL	315	260 320	" 7 " 5
041 213	CALL	315	260 140	Do blank scan
041 216	CALL	315	330 160	Scan dot row 1, character line 6
041 221	CALL	315	330 220	" 3 " 6
041 224	CALL	315	330 260	" 5 " 6
041 227	CALL	315	330 320	" 7 " 6
041 232	CALL	315	330 140	Do blank scan
041 235	CALL	315	010 161	Scan dot row 1, character line 7
041 240	CALL	315	010 221	" 3 " 7
041 243	CALL	315	010 261	" 5 " 7
041 246	CALL	315	010 321	" 7 " 7
041 251	CALL	315	010 141	Do blank scan
041 254	CALL	315	060 161	Scan dot row 1, character line 8
041 257	CALL	315	060 221	" 3 " 8
041 262	CALL	315	060 261	" 5 " 8
041 265	CALL	315	060 321	" 7 " 8
041 270	CALL	315	060 141	Do blank scan
041 273	CALL	315	130 161	Scan dot row 1, character line 9
041 276	CALL	315	130 221	" 3 " 9
041 301	CALL	315	130 261	" 5 " 9
041 304	CALL	315	130 321	" 7 " 9
041 307	CALL	315	130 141	Do blank scan
041 312	CALL	315	210 161	Scan dot row 1, character line 10
041 315	CALL	315	210 221	" 3 " 10
041 320	CALL	315	210 261	" 5 " 10
041 323	CALL	315	210 321	" 7 " 10
041 326	CALL	315	210 141	Do blank scan
041 331	CALL	315	260 161	Scan dot row 1, character line 11
041 334	CALL	315	260 221	" 3 " 11
041 337	CALL	315	260 261	" 5 " 11
041 342	CALL	315	260 321	" 7 " 11
041 345	CALL	315	260 141	Do blank scan
041 350	CALL	315	330 161	Scan dot row 1, character line 12
041 353	CALL	315	330 221	" 3 " 12
041 356	CALL	315	330 261	" 5 " 12
041 361	CALL	315	330 321	" 7 " 12
041 364	CALL	315	330 141	Do blank scan
041 367	LDA	072	000 340	Output //VERTICAL SYNC// pulse
041 372	MVIA	076	175	Load # of vertical blank scans
041 374	CALL	315	015 140	Do //BLANK VERTICAL SCANS//
041 377	DCRA	075		One less blank scan
042 000	MOVAA	177		Equalize 2.5 microseconds
042 001	JNZ	302	(374)(041)	Repeat blank scans if not done
042 004	MOVAA	177		Equalize 5 microseconds
042 005	MOVAA	177		continued
042 006	DI	363		Shut off horn
042 007	JMP	303	(100)(040)	

Notes:

- TVT 6-5/8 must be connected via an 8080 adapter, and both the 658-HD8 and 658-HS80 PROMs must be in circuit for the program to run.
- Address switches must be in "64"; A5="A5"; and A4="A4" positions.
- Horizontal frequency = 11.191 kHz Vertical frequency = 60.006 Hertz. 166 second hum bar.
- This program is not self-modifying and may be placed in PROM or ROM.
- Character sequences goes 340 000; 350 000; 340 001; 350 001; 340 002; 350 002; 340 003.....
- () denotes an absolute address that is program location sensitive.

FLOWCHART:

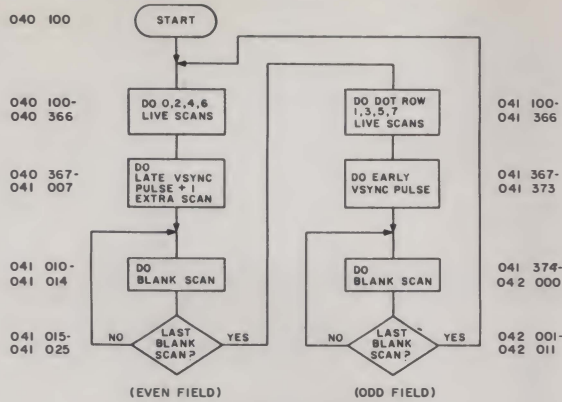


Fig. 14a. Program flowchart.

of set could sing objectionably.

12 Lines of 80 Characters

A brute force, interlaced, double-stuffed 12 x 80 scan program appears in Fig. 14. You can easily modify it for 24 x 80 or even 36 x 80 displays if you like. With the double stuffing, the 12 x 80 display uses slightly less than one-third of the H8 throughput time. By going to suitable transparency techniques, you can save two-thirds of the computer time to transparently run other programs such as Extended BASIC.

We've shown you this scan program with its memory space at 340 010 to 343 377. This assumes you have at least two RAM cards in your H8 and have put this particular one "out on top" with the "56K" jumper on the memory card. You may want to relocate things later, but this is a handy place to start.

The TVT 6-5/8 is attached to the memory card by way of an 8080 adapter similar to Figs. 4 and 10. The TVT does place certain use restrictions on the 340 000 to 360 000 computer address space, since any activity here also gives you a vertical sync pulse that might disrupt an enabled display. You can use this space for a display memory RAM; you should not use this area for the scan program or the computer stack. If you do use this page for display memory RAM, you will have to watch your cursor program carefully if transparent character entry is important to you.

You'll find the 12 x 80 program shown in two separate fields. We have an even field that puts down the even dot rows of all the characters and an odd field that puts down the odd dot rows of all the characters. When combined, these fields form an interlaced and double-stuffed frame. Having the two fields separate is handy for debugging. By jumping a field back on itself, you can display all-even or all-odd fields to fix coding errors or make format changes.

The scan program runs just about the same way the earlier 1 x 56 program did. First, there is a short blank scan; then we put down the even dot rows of all the characters. Then we equalize, followed by a late vertical sync pulse, at the same time taking up one entire extra horizontal scan time. Then we run the usual blank vertical scans, completing the field.

When the field is finished, we jump to the odd field, run a short blank scan and put down all the odd dot rows of all the characters. After this, we run an early vertical sync pulse and go on to the usual number of vertical blank scans. The scan sequence repeats by jumping to the start of an even field.

The early and late vertical sync pulses differ by half a horizontal line. When you combine this half a line with the extra horizontal line picked up only in the even scan, you end up with an interlaced scan of 373 whole lines taking one 30 Hz frame. This 30 Hz frame consists of

two 60 Hz fields of 186.5 lines each.

The 658-HS80 Scan PROM lets you repack the 80 character lines so you can use your display memory space efficiently. Fig. 15 shows how the characters are arranged in RAM. While this looks like a royal mess, a few extra cursor words are all we need to straighten things out. This is often a reasonable trade-off for letting us do long lines with an 8080 in the first place and freeing up 600 or so words of system RAM for other uses.

Your turn: Show the coding for 24 x 80, 32 x 80, 16 x 56, 32 x 56, 16 x 64 and 32 x 64 scan programs. Show ways of significantly shortening the 12 x 80 scan program while staying

PROM compatible. Try: (1) using only one vertical blanking sequence and minimizing blank sequences and unused code words; (2) using I/O commands to free address space; (3) using interrupt rather than subroutine mapping.

Note that you'll use the HS64 PROM for 64 and shorter character lines and most graphics, while the HS80 PROM is usually reserved for 80 character lines. You can do 40 character lines with the HS64 without repacking, or else you can use your memory more efficiently by going to a specially coded HS40 PROM that uses repacking. Repacking saves you RAM space but needs a few extra words in the cursor program and takes a special Scan

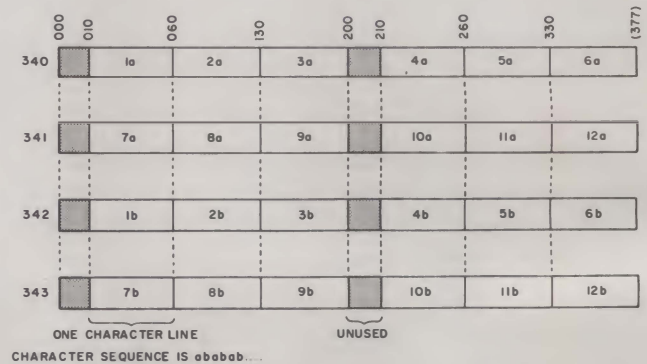


Fig. 15. Display memory map for 12 x 80 scan.

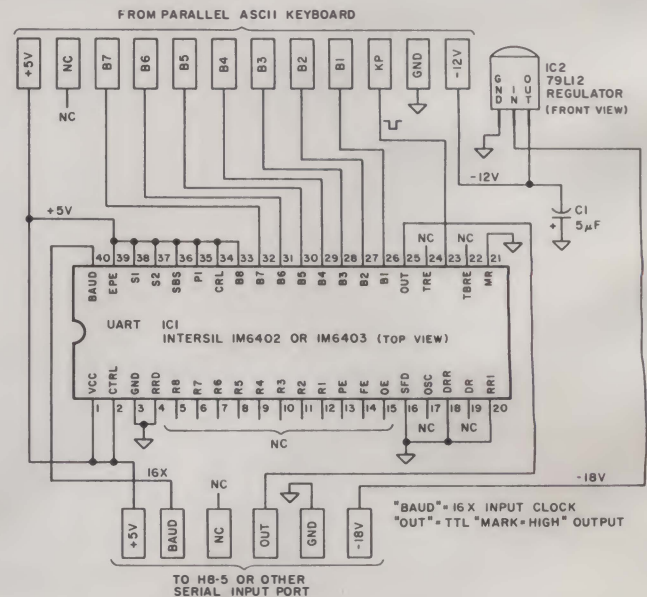


Fig. 16. This keyboard serial adapter lets you connect a keyboard to a serial computer input.

PROM.

A Keyboard Serial Adapter

If you have an H8-2 parallel interface card, it should be fairly easy to interface almost any old ASCII keyboard and encoder. You could do this essentially the same way we did it on the parallel KIM inputs back in

the *Cheap Video Cookbook*, but the H8-2 card is an expensive option and you might not already have one on hand. More likely, you'll be using the H8-5 serial interface card instead, since you need this one for the usual cassette and remote terminal uses.

Most ASCII keyboards and

encoders provide only a parallel (all-the-bits-at-once) output. To enter a serial port, we have to convert this parallel word into a serial (one-bit-at-a-time) sequence. A simple adapter to do this is shown in Fig. 16.

The circuit can use the transmitter half of nearly any old UART (universal asynchronous receiver-transmitter). We first looked at UARTs back in Chapter 7 of the *TVT Cookbook*. You'll find this circuit easier and more inexpensive when you use a modern, single-supply CMOS chip such as an *Intersil* iM6402 or iM6403.

The keyboard serial adapter works by borrowing power from the H8-5 serial interface and feeding +5 volts and optionally -12 volts to your existing keyboard. Your existing keyboard outputs are most likely available in *parallel* or "all-at-once" form. These parallel outputs and a normally-high keypressed *strobe* are routed to the transmitter side of the UART in the adapter. This UART also borrows a 16X baud clock from the H8-5.

When you press a key, a serial output is generated by the UART. This serial output is then routed to your computer's serial interface and received just as if it came from a terminal.

You may need as many as five leads between your adapter and the H8-5. One is ground, two are for power, one is the 16X baud rate clock that goes to the adapter and the final is the serial output that comes from the adapter. Fig. 17 shows

you how to connect, both pictorially and schematically, your adapter to your H8-5. You can either hard-wire these connections or add a new connector of your own.

On your H8-5 board, integrated circuit IC122 is removed and replaced with two jumpers inserted in the socket as shown. The pin-11-to-pin-13 jumper gives you direct access to the serial input on the UART present inside the H8-5. The pin-6-to-pin-7 jumper lets you use the keyboard in a *polled* mode. This polled operation gives you a transparent scan program and frees the interrupts for other uses.

The H8 has to be software-programmed to use your new adapter. A simple test sequence that will enter the last-pressed key into the accumulator and display it for you is shown in Fig. 18.

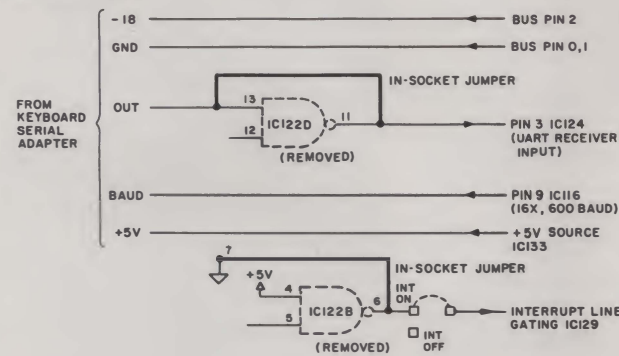
The H8-5 is first initialized with a *mode* instruction. You can use 312 and output it to port 373. This picks two stop bits, ignores parity, uses a seven-bit word and runs with a 16X clock. Next, you continue to initialize the H8-5 by giving a *command* instruction to the same port. This time, use 004 and once again output it to port 373. This command instruction will enable only the receiver in the H8-5 interface.

After the mode instruction and the command instruction are routed to the interface, you are free to read characters. You do this by inputting from port 372. The final loop in the test program does this continuously.

As you press a key, its ASCII value will appear in the left three digits of the "AF" Register display. For instance, a lowercase "b" will read 142, while an uppercase "B" will read 102.

There are a few gotchas in this simple test program, so you'll want to improve it for actual use as part of a cursor. Note that this simple program continuously *rereads* characters instead of reading each one just once. To beat this, there is available a "character ready" (R x RDY) flag that is set when

(a) SCHEMATIC



(b) PICTORIAL

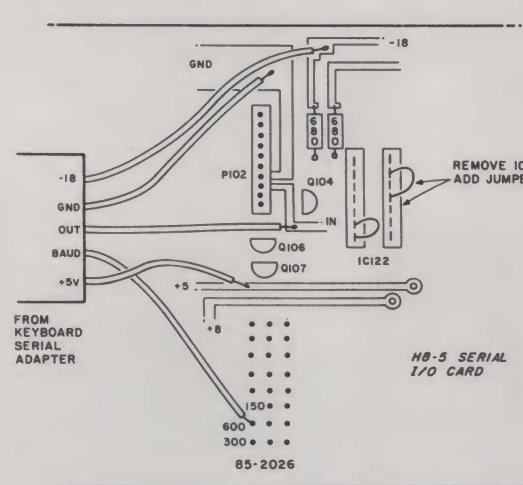


Fig. 17. Connecting your keyboard serial adapter to an H8-5 interface.

uP-8080A		System-H8 + H8/5		Program Space		Start-JMP 040 100	End-RST/0
START	040 100	MVIA	076 312			Initialize mode instruction	
	040 102	OUT	323 373			continued	
	040 104	MVIA	076 004			Initialize command instruction	
	040 106	OUT	323 373			continued	
	040 110	IN	333 372			Read Keyboard	
	040 112	JMP	303 (110)(040)			Loop	

Notes:

- This test program displays a pressed key received via the Keyboard Serial Adapter. To run the program, use: RST/0-REG-PC-ALTER-0-4-0-10-0-ALTER-REG-AF-GO.
- ASCII characters should appear as the three leftmost digits on the display. For instance, "A" = 101, "a" = 141, "6" = 066, and "CR" = 015.
- () Denotes an absolute address that is relocation sensitive.

Fig. 18. Keyboard serial adapter test program.

the character first arrives and is reset as soon as the computer uses the character for the first time.

To use a character only once, input from port 373, AND what you get with 002 and test the result. A nonzero result means you have a new character ready to enter. A zero result says you have already used the character on-hand and should ignore it. We'll see an example on this shortly.

The UART doing the transmitting (in the adapter) and the one doing the receiving (in the H8-5) must agree on the baud rate and the baud clock factor. Usually, the H8-5 will be set on 600 baud and 16X clocks with internal jumpers. If not, or if you are on a different system, be sure that the transmitting UART and the receiving UART are on speaking terms with each other.

Note that your initialization of the *mode* and *command* words should be done only once after reset and before any input/output activity. If you don't initialize, you'll get no characters at all, and if you continuously re-initialize, characters will get dumped before you can use them.

Your keyboard serial adapter is very flexible. For instance, go over the data sheets to find a whole unused UART receiver on the low number pins. The -12 volt supply is an option. You can eliminate it if you already have -12 on hand or use a keyboard that doesn't need it. You can also use the old-style UARTs that need -12 by removing the connections on pin #2 and jumpering to -12.

Should you use the IM6403, you can eliminate the 16X baud rate line by connecting a 3.58 MHz color TV crystal between pins 17 and 40 while grounding pin 3 of the IM6403. This will output characters for you at 110 baud. Your computer's serial input will also have to be jumpered or programmed to use this new data rate.

As shown, the keyboard serial adapter is programmed to provide a permanent one in the transmitted ASCII bit #8, is continuously enabled, has no parity,

uses two stop bits and has an eight-bit word length. You can change any or all of these by re-programming the hard-foil connections of pins 33 through 39 of the UART. Our circuit assumes the keyboard outputs positive logic and uses a narrow goes-to-ground-from-positive-high strobe that is low only when data is valid. The output is a simple TTL logic level. There is no need to convert to

RS-232 or Teletype current loops for a short interface connection.

Your turn: Show how to use your keyboard serial adapter with only two wires between computer and keyboard, including all power supply connections. Hint: Use the IM6403 with a crystal and a CMOS-encoded keyboard. Change the current when you want to send a zero and sense this current at the

computer end.

If you really want to get fancy, use ultrasonic or infrared transducers to give you zero connections between keyboard and computer. This will, of course, take batteries inside the keyboard, or will it?

8080 Cursor Software

Many of the ideas we have already used for our previous cur-

Fig. 19. Program for a one-line, 56-character TVT 6-5/8 8080 raster scan integrated minimum cursor.

	uP-8080A	Start-JMP 040 100	Displayed	340 004 to 340 037
	System-H8	End-RST/0		342 004 to 342 037
			Program Space	040 100 to 040 341
			Registers Used	-B, H, L

Main scan sequence:

START →	040 100	MVIA	076	312	Initialize MODE for H8-3
	040 102	OUT	323	373	continued
	040 104	MVIA	076	004	Initialize COMMAND for H8-3
	040 106	OUT	323	373	continued
	040 110	CALL	315	017 140	Do short BLANK SCAN
	040 113	CALL	315	004 160	Scan Dot row #1
	040 116	CALL	315	004 200	Scan Dot row #2
	040 121	CALL	315	004 220	Scan Dot row #3
	040 124	CALL	315	004 240	Scan Dot row #4
	040 127	CALL	315	004 260	Scan Dot row #5
	040 132	CALL	315	004 300	Scan Dot row #6
	040 135	CALL	315	004 320	Scan Dot row #7
	040 140	MVIB	006	364	Load number of blank scans in B
	040 142	IN	333	373	Is a new key pressed?
	040 144	ANI	346	002	Mask depressed bit
	040 146	JZ	312	(154)(040)	No, continue scan
040 220 →	040 151	CALL	315	(220)(040)	Yes, go to cursor
	040 154	CALL	315	015 140	Do equalizing BLANK SCAN
	040 157	LDA	072	000 340	Output vertical sync pulse
	040 162	MOVBA	170	170	Get number of blank scans back
	040 164	CALL	315	011 140	Do BLANK SCAN
	040 167	DCRA	075		One less scan
	040 170	MOVAA	177		Equalize 2.5 microseconds
	040 171	JNZ	302	(164)(040)	Do another blank scan?
	040 174	MOVAA	177	177	Equalize 5 microseconds
	040 176	DI	363		Shut Off Horn
	040 177	JMP	303	(110)(040)	Go to new field

CURSOR RETURN

Cursor Processing Subroutine:

040 151 →	040 220	MOVAH	174		Get upper cursor address
(Enter)	040 221	ANI	346	375	Mask A9 out
	040 223	CPI	376	340	Is upper page address valid?
	040 225	JZ	312	(233)(040)	Yes, OK to continue
040 260 →	040 230	CALL	315	(260)(040)	No, clear screen via subroutine
	040 233	MOVAL	175		Get lower cursor address
	040 234	ANI	346	037	Put it on the screen
	040 236	MOVLA	157		Replace lower cursor
	040 237	IN	333	372	Get character
	040 241	CPI	376	015	Is it Carriage Return (Erase)?
	040 243	JZ	312	(260)(040)	Yes, clear screen via subroutine
040 300 →	040 246	CALL	315	(300)(040)	No, enter character via subroutine
	040 251	RET	311		Return to scan program

(040 251 through 040 257 are spares; not used)

040 320 →	040 260	CALL	315	(320)(040)	go to clear screen subroutine
	040 263	MVIB	006	331	Equalize # of blank scans remaining
040 154 →	040 265	RET	311		Return to Processing

(Exit)

Enter Character and Increment Subroutine:

```

enter → 040 300 MOVMA 167      Store character at cursed location
        040 301 MOVVA 174      Get upper cursor word
        040 302 XRI 356 002     Change address A9
        040 304 MOVVA 147      Replace upper cursor word

        040 305 ANI 346 002     Is address A9 now zero?
        040 307 RNZ 300         No, return
        040 310 INXH 043        Yes, increment HL (cursor address)
        040 311 RET 311        Return to Processing

Exit 1 → 040 320 LXIH 041 (004)(340) Home Cursor
        040 323 MVIA 076 040    Load Space
        040 325 CALL 315 (300)(040) Enter space via ECI subroutine
        040 330 MVIA 076 040    Is it the end of the screen?

        040 332 CMPL 275         continued...
        040 333 JNC 302 (323)(040) No, add more spaces
        040 336 LXIH 041 (004)(340) Yes, home cursor
Exit 2 → 040 341 RET 311        Return to Processing
    
```

Notes:

- TVT6-5/8 must be connected via an 8080 adapter and both the 658-HD8 and 658-HS64 PROMs must be in circuit for the program to run. Character entry via keyboard, a keyboard serial adapter and the H8-3 serial interface card.
- All characters and all control commands are entered on the screen, except for carriage return (CR), which clears the screen.
- Horizontal frequency is 15.174 kHz; Vertical frequency is 59.976 Hz. 2500 second hum bar.
- Address switches must be in "32"; A5 = "+"; and A4 = "A4" positions.
- Character sequence goes 340 004; 342 004; 340 005; 342 005; 340 006; 342 006; 340 007. . .
- This program is not self-modifying and may be placed in PROM or ROM. Register "B" is used for temporary storage; Registers "HL" are used to hold the cursor address.
- To aid in debugging, replace 040 147 with 000 and manually defeat front panel interrupt. To shorten number of characters displayed for a tv with limited width, use 040 337 value of 005 or higher.
- () denotes an absolute address that is program location sensitive.

sors will carry over to 8080 cursor design. One new hassle we'll pick up is the straightening-out process needed to undo the A9 speed doubling. But this is more than offset by the easier and simpler code using all the available 8080 registers, particularly the 16-bit wide HL register that is ideal for cursor location storage.

Let's look at a simple cursor that ties the keyboard input to an 8080 display. We'll use the 1x56 display to keep things simple. The program and a flowchart are shown in Fig. 19.

For convenience, we've left this program in several pieces, omitted a visible cursor and done only "good enough" equalization. While you can use this program for a one-line point-of-sale terminal, as a deaf communicator or in a prompting environment, chances are that you'll want to pick up these bits and pieces and then combine them with the best of the earlier cursors to do your own thing.

Our main scan sequence is about the same as the old 1x56 scan program of Fig. 13. We've added some words at the start that initialize our H8-5 serial interface so it will accept a keyboard input by way of the keyboard serial adapter. Our brute force scans are called for next

as needed to give us a line of characters.

After the characters are down, we test to see if a new key has been pressed. If not, we output a vertical sync pulse, run the blank vertical retrace scans, and then jump up and repeat everything for the next field. Note that we do *not* re-initialize the serial interface each time. We simply loop back to the start of the next field.

Now, if a key has been pressed, we jump to the new *Cursor Processing* subroutine at 040 220 through 040 251. This cursor processing subroutine first checks to make sure the HL register is holding a valid cursor location. If it isn't, the screen is erased and the cursor fixed before anything happens to other programs in the machine.

We then get a character and test it to see if it is a CR, or carriage return. If it is a CR, we erase the screen and home the cursor. CR was chosen over CAN in this example as it seems more appropriate for a one-line display. You can, of course, use any decoding you like.

If any key *but* the carriage return is pressed, the character is entered. This is done by way of an enter-character-and-incre-

ment, or ECI, subroutine. This ECI subroutine is fancier than the ones we used before, since we have the A9 switching to contend with. Some new rules and a few extra code words take care of this for us.

Remember that the A9 switching was used to let us get characters out of the 8080 fast enough to be useful. To do this, the display characters are out of order. Specifically, for our 1x56 display, the character sequence goes like this:

1st character 340 004

2nd character 342 004
 3rd character 340 005
 4th character 342 005

55th character 340 037
 56th character 342 037

Now every time we enter a character, we want to go on to the next one. So, we first *change* A9. To do this, we use an exclusive OR 002 of the H register. This will automatically make A9 a one for a particularly character, a zero for the next character, a one for yet the next character and so on.

If A9 goes from a zero to a one, we need do nothing further. If A9 goes from a one to a zero, however, we need to move onto the next pair of character slots in memory. To do this, we increment the HL register that contains the cursor.

So, we change A9 *every* new character but increment our HL cursor only every *second* character. All the A9 switching mess is magically eliminated with nothing but eight or so program words.

Your turn: Show an all-the-bells-and-whistles cursor for a 24x80 display, including a visible cursor, full equalization and transparency, all cursor motions and the usual goodies.

As with the 6502 systems, there is virtually no limit to how fancy your cursor programs can become. All it takes are extra words of machine-language code to do almost anything you can dream up. ■

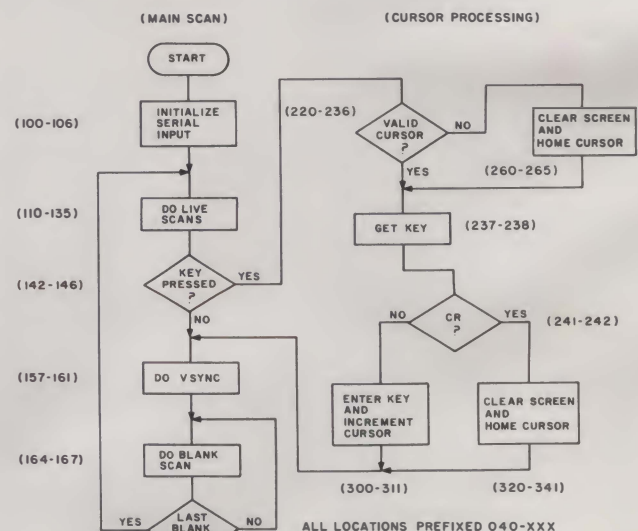


Fig. 19a. Program flowchart.